# **FREQUENCY COMPENSATION**

#### INTRODUCTION

In this lecture, we deal with the stability and frequency compensation of linear feedback systems to the extent necessary to understand design issues of analog feedback circuits. Beginning with a review of stability criteria and the concept of phase margin, we study frequency compensation, introducing various techniques suited to different op amp topologies. We also analyze the impact of frequency compensation on the slew rate of two- stage op amps.

## **BASIC NEGATIVE-FEEDBACK SYSTEM**

General considerations

□ Close-loop transfer function:  $\frac{Y}{X}(s) = \frac{H(s)}{1 + \beta H(s)}$ if  $\beta H(s = j\omega 1) = -1$ , the gain goes to infinity, and the circuit can amplify is own noise until it eventually begins to oscillate at frequency  $\omega 1$ .

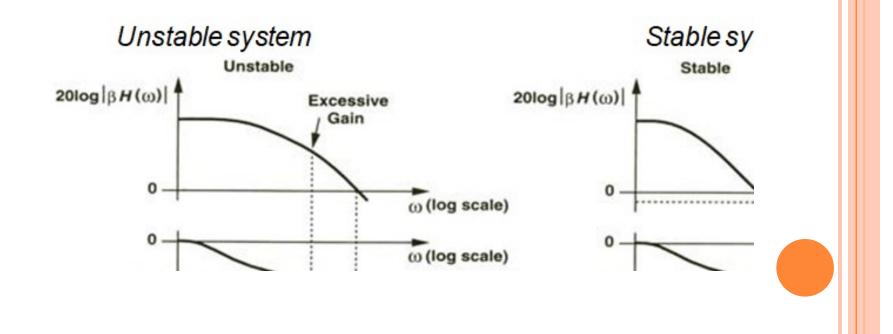
Barkhausen's Criteria:

 $|\beta H(s=j\omega 1)| = 1$  $\angle \beta H(s=j\omega 1) = -180^{\circ}.$ 

The total phase shift around the loop at  $\omega_1$  is 360° becaus feedback itself introduces 180° of phase shift. The 360° phase shift around the loop at  $\omega_1$  is 360° becaus

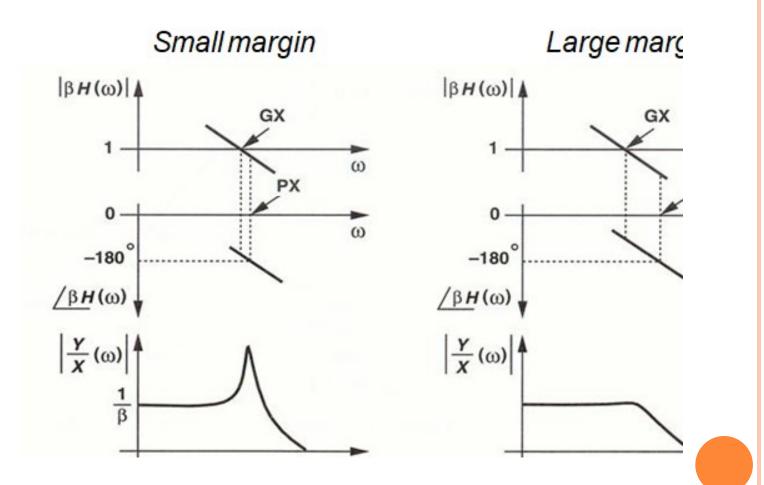
# BODE PLOT

- A negative feedback system may oscillate at ω1 if
  - the phase shift around the loop at this frequency is so mu feedback becomes positive.
  - (2) the loop gain is still enough to allow signal buildup.



# Phase margin

Close-loop frequency and time response



#### Phase margin (cont'd)

• Phase margin (PM) is defined as

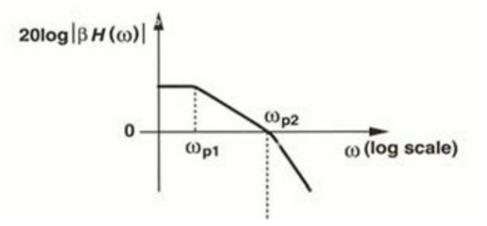
$$PM = 180^{\circ} + \angle \beta H(\omega = \omega_1)$$

Since  $\angle \beta H$  reache

where  $\omega_1$  is the gain crossover frequency.

Example

A two-pole feedback system is designed such that  $|\beta H(\omega = \alpha |\omega_{P1}| \le |\omega_{P2}|$ .



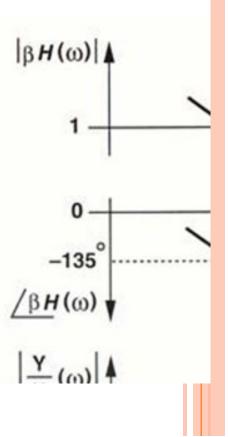
#### How much phase margin is adequate?

• For  $PM = 45^\circ$ , at the gain crossover frequency  $\angle \beta H(\omega_1) = -13^\circ$  $|\beta H(\omega_1)| = 1$ , yielding  $\underline{Y} = H(j\omega_1) = -13^\circ$  $X1 + 1 \cdot \exp(-j135^\circ) = 0.29 - 0.71j$ 

It follows that  $\left| \frac{Y}{X} \right| = \frac{1}{\beta} \cdot \frac{1}{\left| 0.29 - 0.71j \right|} \beta$ 

The frequency response of the feedback system suffers from a 30% peak at  $\omega = \omega_1$ .

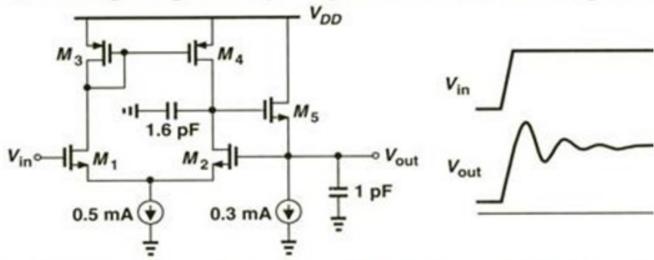
 Close-loop frequency response for 45° phase margin:



### How much phase margin is adequate? (cont'd)

Example:

Unity-gain buffer:  $PM \approx 65^{\circ}$ , unity-gain frequency = 150 MHz. However, the large-signal step response suffers from significations of the step response suffers from step response suffers



 The large-signal step response of feedback amplifiers is n slewing but also because of the nonlinear behavior result excursions in the bias voltages and currents of the a excursions in fact cause the note and zero frequencies to p